

# Automated Reconfigurable Antenna Impedance for Optimum Power Transfer

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**Abstract:** This paper presents an approach to implement an automatically tuning antenna for optimising power transfer suitable for software defined radio (SDR). The automatic tunability of the frequency selective antenna is achieved by a closed-loop automatic antenna tuning network comprising of an impedance sensor and control unit. The sensor provides the control unit with data on the transmit or receive power, and the an algorithm is used to impedance of a  $T$ -network of  $LC$  components to optimize the antenna impedance to maximise power transmission or reception. The effectiveness of the proposed tuning algorithm in terms of impedance matching and convergence on the optimum matching network values is shown to be superior compared with the conventional tuning algorithm.

**Keywords:** Impedance matching, Antenna, Automatic impedance tuning, Impedance matching algorithm, power transfer, transmitter power amplifier (PA), low-noise amplifier (LNA) receiver.

## I. INTRODUCTION

In wireless communication systems the function of an antenna is to provide an interface with free-space for transmission and reception of the EM signal. Antennas therefore have an impact on the performance of the wireless system in terms of efficiency and radiation pattern [1]. In fact, at the transmitter the antenna is the load impedance for the power amplifier (PA), and the source impedance for the low-noise amplifier (LNA) receiver. The feed-point impedance therefore differs widely for optimum PA efficiency or LNA noise performance. Furthermore, the impedance also varies with frequency and with environmental effects, for example the proximity of a portable wireless device to the user's body.

Wireless communication systems are evolving fast with ever decreasing size, increasing bandwidth and convergence of applications within one device. In fact, the requirement is to use a single wideband antenna that can

accommodate several communications standards to facilitate software-defined operation.

To circumvent the feed-point impedance issue requires incorporation of a tuneable impedance matching network between the transceiver and the antenna. In this paper an automatic antenna tuning network (AATN) is presented for integration in wireless communication systems.

## II. IMPEDANCE MATCHING AND TUNING

The purpose of the automatic antenna tuning network, shown in Fig.1, is to maximise the transfer of power between transmitter and antenna and maximise the signal-to-noise ratio of the received signal. This can be achieved by matching the impedance of the antenna ( $Z_A$ ) over the operating frequency range to the load impedance ( $Z_L$ ) [2]-[4].

AATN comprises (i) a tuneable matching system to transform the antenna's impedance for optimum impedance matching; (ii) an impedance sensor block; and (iii) a control unit that uses the information from the sensor to modify the  $LC$  values of the matching network to optimise the impedance matching to the antenna.

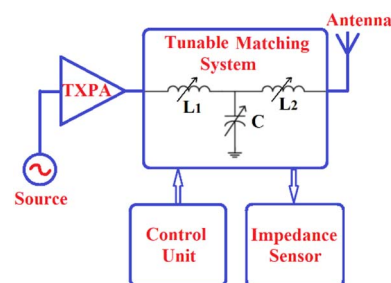


Fig.1. Block diagram of the automatic antenna tuning network with  $T$ -tuneable matching system.

## III. MATCHING NETWORK

For optimum power transfer a  $T$ -network matching circuit in Fig.2 is employed for impedance transformation

between the antenna and (i) transmitter PA; and (ii) receiver LNA. This low-pass filter topology offers the advantage of minimizing noise in the system.

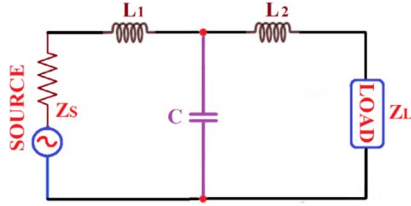


Fig.2. Schematic of the  $T$ -network matching circuit model.

Application of simple circuit analysis reveals the feasibility of the  $T$ -network to transform the source impedance to the load impedance determined by the networks  $Q_0$  value. In the analysis it's assumed that the source impedance and load impedance are both resistive, i.e.  $Z_S = R_1$  and  $Z_L = R_2$ , so:

$$X_{L1} = R_1 Q_0 \quad (1)$$

$$X_{L2} = \frac{1}{R_2} \sqrt{\frac{Q_0 + 1 - R_1 R_2}{R_1 R_2}} \quad (2)$$

$$X_C = \frac{R_1(Q_0 + 1)}{Q_0 + X_{L2} R_2} \quad (3)$$

where  $X_{L1}$ ,  $X_{L2}$ , and  $X_C$  are the reactance of  $L_1$ ,  $L_2$ , and  $C$ , respectively at the operating frequency. Quality factor is given by:

$$Q_0 \geq \sqrt{\frac{R_2}{R_1}} - 1 \quad (4)$$

The above analysis can be extended to the case where  $Z_S$  and  $Z_L$  are complex [5].

Fig.3 shows the how the tuneable  $C$  and  $L$  components in the  $T$ -network can be implemented in practice using a switched array of fixed values. It is evident that with this approach there are practical limitations on the range of impedance values for matching. Hence, a given matching network will provide a finite range of matching impedances. Therefore, antenna impedances that fall outside this range cannot be matched. In addition, the minimum incremental values of  $C$  or  $L$  in the network are limited by circuit parasitics. For an on-chip application the switched series inductor array is not feasible because it would require an excessive chip footprint. In addition, the series connection of the switches in the array will lead to a high switch resistance that would increase loss.

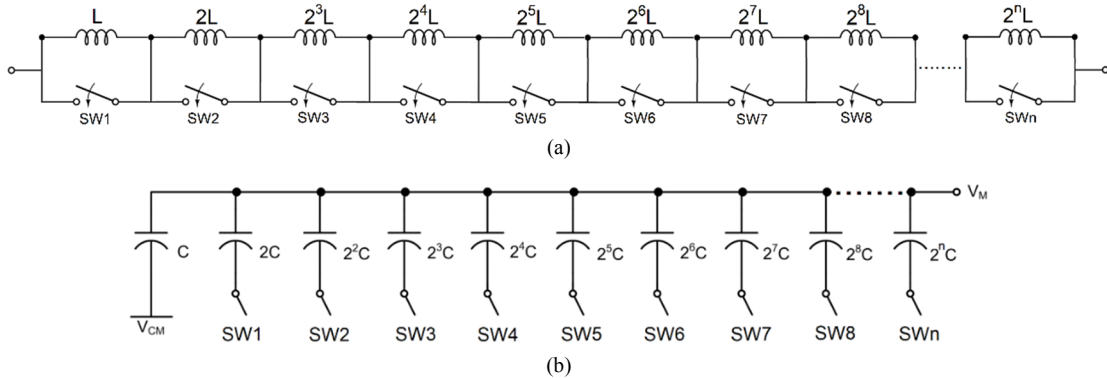


Fig.3. Switchable arrays of (a) series inductors, and (b) shunt capacitors.

For limited bandwidth applications, the  $T$ -network with series tuneable inductor (STI) can be replaced with a shunt tuneable capacitor (STC) in conjunction with impedance inverting networks (IINs) as shown in Fig.4. This topology can be applied for CMOS on-chip antenna matching, as shown in Fig.5 [6, 7], where the active device is configured as shunt switch. This approach minimises the switch on-resistance and reduces production of intermodulation products due to switch non-linearity.

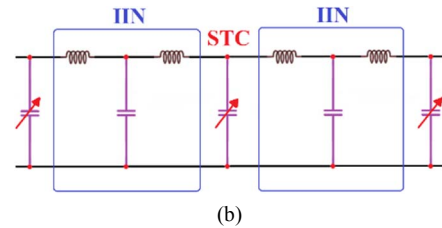
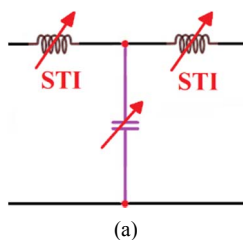


Fig.4. (a)  $T$ -network circuit model with series tuneable inductor (STI), and (b) Transformed matching network with shunt tuning capacitor (STC) in conjunction with impedance inverting networks (IINs).

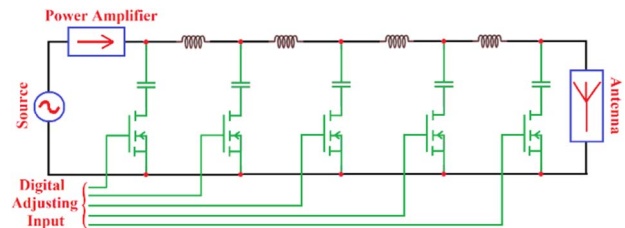


Fig.5.  $T$ -network circuit model for CMOS on-chip antenna impedance matching.

Performance of the switches themselves can have a great impact on the impedance tuning network. In fact, the IC fabrication technology will determine the type of switches to be used. Switches implemented with active devices can introduce extra loss and circuit parasitics as well as unwanted harmonics and intermodulation products. Their limited breakdown voltages limit the power-handling capability of the matching network.

#### IV. IMPEDANCE SENSORS

The purpose of the impedance sensor in Fig.1 is to provide feedback to the tuning control unit so that the antenna's impedance is appropriately matched for optimum power transfer. The sensor essentially detects the power of the transmitted signal at the antenna terminals. The PA may operate with compromised efficiency as maximum power condition does not necessarily coincide with the optimum load impedance. Output voltage and current from the PA can also be monitored using a phase detector. This information can be used by the control unit to modify the matching network for minimum phase differential; thus, making the load impedance at the PA output resistive for improving power transfer [8].

#### V. TUNING ALGORITHM

In a tuning network with a relatively small number of  $LC$  components it is possible to perform an extensive search of all possible combinations in order to find the one producing the optimum impedance match. However, this is not the case for a larger network. One method to obtain this is to generate a look-up table of matching network data for various operating frequencies during an initialisation phase; the control unit then chooses the proper adjusting data from the table as the working frequency is changed. This approach obtains quick adjusting but is not able of responding to variations in antenna impedance that happen over time without repeating the initialisation process. Here a tuning algorithm has been developed that uses impedance sensor data to iteratively converge on the optimum matching network values. Rapid convergence is achieved because the sensor provides both signal amplitude and phase information.

To confirm the effectiveness of the proposed tuning algorithm (PTA) it is compared with the conventional tuning algorithm (CTA). In CTA as the sensors provide only amplitude information the algorithm uses a "trial-and-error" approach to determine the impedance matching mismatch. The tuning times of the algorithms was

assessed over 400 simulations to determine the algorithm's speed in locating the optimal solution. For source impedance of  $Z_S = 50 + j20 \Omega$ , load impedance of  $Z_L = 20 + j30 \Omega$ , source signal frequency of 5 GHz (WiFi band), and signal amplitude of 1 V. The optimum solution obtained gives  $L_1 = 5.62 \text{ nH}$ ,  $L_2 = 3.45 \text{ nH}$ , and  $C = 2.18 \text{ pF}$ . The elapsed time of PTA was 4.25s and CTA was 10.54s. This demonstrates the effectiveness of the proposed solution.

#### VI. CONCLUSIONS

In this paper an effective technique is described for automatically tuning the impedance of an antenna for integration in wireless communication systems. The proposed technique minimizes the number of iterations needed to complete the tuning process. Compared with conventional tuning techniques the proposed algorithm rapidly converges to an optimum or close to the optimum impedance for maximum power transfer.

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#### REFERENCES

- [1] J. D. Kraus and R. J. Marhefka, *Antennas*, McGraw-Hill, NY, 2002.
- [2] J. R. Moritz and Y. Sun, "Frequency agile antenna tuning and matching," in Proc. Eighth International IEE Conference on HF Radio Systems and Techniques, pp. 169 -174, Surrey, UK, July 2000.
- [3] Yichuang Sun, James Moritz, and Xi Zhu, "Adaptive impedance matching and antenna tuning for green software-defined and cognitive radio", 2011 IEEE 54th International Midwest Symposium on Circuits and Systems (MWSCAS), Seoul, South Korea, pp. 7-10 Aug. 2011.
- [4] J. T. Aberle, B. Bakkaloglu, C. Chakrabarti, S. -H. Oh, G. A. Taylor, H. Song, A. Adhya, K. L. Melde, R. B. Whatley, and Z. Zhou, "Automatically tuning antenna for software-defined and cognitive radio," Software Defined Radio Technical Conference and Product Exposition SDR, Session 5.4, 6 pages, California, USA, Nov. 2005.
- [5] Y. Sun and J. K. Fidler, "Design method for impedance matching networks", IEE Proc. Circuits, Devices and Systems, Vol. 143, No. 4, pp 186-194, August 1996.
- [6] A. Chamseddine, J. W. Haslett, and M. Okoniewski, "CMOS silicon-on-sapphire tunable matching networks", EURASIP Journal on Wireless Communications and Networking, Vol. 2006, Article ID 86531, 2006.
- [7] P. Sjoblom and H. Sjoland, "An adaptive impedance tuning CMOS circuit for ISM 2.4-GHz band", IEEE Transactions on Circuits and Systems-I, Vol. 52, No. 6, pp.1125 -1124, June 2005.
- [8] A. Zolomy, F. Mernyei, J. Erdelyi, M. Pardoen, and G. Toth, "Automatic antenna tuning for RF transmitter IC applying high Q antenna", IEEE Radio Frequency Integrated Circuits Symposium, paper TU4D-4, pp. 501-504, June 2004.